

## PhD Candidacy Exam - Signal Processing

1. Given a continuous-time signal  $x(t)$  with  $X^F(\omega) = 0$  for  $|\omega| > \omega_m$  determine the minimum sampling rate  $f_s$  for a signal  $y(t)$  defined by

(a)  $x^2(t)$

(b)  $x(2t)$

(c)  $x(t) \cos(6\pi\omega_m t)$

2. (a) Find the period of the signal  $f[k] = e^{j\frac{\pi}{16}k} \cos\left(\frac{\pi}{17}k\right)$ .  
(b) Find the energy and power of the signal  $f[k] = (1.5)^k u[-k]$ .  
(c) For a DT system with input  $f[k]$  and output  $y[k]$ , if  $y[k] = f[k] \sin\left(\frac{\pi}{2}k\right)$ , determine whether the system is *linear time invariant (LTI)* or not.  
(d) Determine whether the system defined by  $y[k] = f[k] * \cos\left(\frac{\pi}{8}k\right)$  BIBO stable or not.
3. (a) A system's response to  $x(n)$  is  $y(n)$  as the followings:

$$x(n) = \left(\frac{1}{2}\right)^n u(n) - \frac{1}{4} \left(\frac{1}{2}\right)^{n-1} u(n-1)$$

$$y(n) = (1/3)^n u(n)$$

- Determine the impulse response  $h(n)$  and the system function  $H(z)$  of a system that satisfies the above condition.
  - Find the difference equation that characterizes this system.
  - Determine if the system is stable.
- (b) Using the properties of  $Z$ -Transform, show that for a real  $x(n)$  that has a  $H(z)$ , if a pole (zero) of  $H(z)$  occurs at  $z = z_0$ , then a pole (zero) must also occur at  $z = z_0^*$ .
4. Consider signal  $x[n] = \{-1, 2, -3, 2, -1\}$  with discrete-time Fourier transform  $X^f(\theta)$ . Compute the following quantities:

(a)  $X^f(0)$

(b)  $\int_{-\infty}^{\infty} X^f(\theta) d\theta$

(c)  $X^f(\pi)$

(d)  $\int_{-\pi}^{\pi} |X^f(\theta)|^2 d\theta$

5. (a) Let  $x(n)$ ,  $n = 0, 1, \dots, 7$  be a purely imaginary sequence. The first five points of an eight-point DFT of this sequence are

$$\{j0.55, -0.055 + j0.1636, 0.3 + j0.15, 1.0450 + j1.4364, -j0.05\}.$$

Determine the remaining three points and explain the basis of your answer.

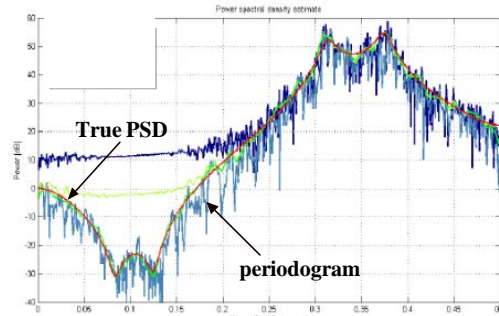
- (b) Let  $y(n)$ ,  $n = 0, \dots, N - 1$  be a complex sequence, whose  $N$ -point DFT is  $Y(k) = \alpha_k$ ,  $k = 0, \dots, N - 1$ . Find an expression for the  $N$ -point DFT of the complex conjugate of  $y(n)$ , in terms of  $\alpha_0, \dots, \alpha_{N-1}$ .
- (c) We wish to obtain the spectrum of an analog signal whose bandwidth is 8 kHz, by first sampling the signal and then computing the DFT. For practical reasons, we have to use the minimum possible sampling frequency. If we desire a spectral resolution of at least 40 Hz, determine the minimum allowable duration of the analog signal record. Clearly explain how you arrive at your result.
- (d) A 7.5 kHz sinusoidal signal is sampled at the rate of 20 kHz, and 256 samples are used to compute the 512-point DFT  $X(k)$ ,  $k = 0, \dots, 511$ . At which  $k$  would you expect to see a peak in  $X(k)$ ? Explain the reasoning behind your answer.
6. Consider the single-pole digital filter given by

$$y(n) = Q_b[ay(n - 1)] + x(n),$$

where  $x(n)$  and  $y(n)$  are the input and output respectively,  $a$  is the filter coefficient and  $Q_b[\cdot]$  denotes the  $b$ -bit round-off operation in the digital-multiplier.

- (a) One simple way to analyze the effect of round-off noise is to represent the round-off operation by the *additive noise model* for quantization errors. State the key assumptions made in this model.
- (b) Using this model, derive a closed-form expression (in terms of  $a$  and  $b$ ) for the variance of the error that will appear at the *output* of the filter. Clearly show how you arrive at the solution.
7. An AR(3) process is characterized by prediction coefficients:  $a_3(1) = -1.25$ ,  $a_3(2) = 1.25$ ,  $a_3(3) = -1$
- (a) Determine its lattice filter reflection coefficients.
- (b) Determine its autocorrelation  $r_{xx}(m)$  for  $0 \leq m \leq 3$ .
- (c) Determine the mean-square prediction error.

8. Let  $\{X(n)\}$  be a wide-sense real-valued stationary stochastic process with  $E\{X(n)\} = 0$ . Let  $r(k) = E\{X(n)X(n-k)\}$ ,  $k = 0, \pm 1, \pm 2, \dots$  denote the autocorrelation function and  $S(f) = \sum_{k=-\infty}^{\infty} r(k)e^{-j2\pi fk}$  the power spectral density (PSD) of the process. Let  $X(0), X(1), \dots, X(N-1)$  be an observed data record of the process  $\{X(n)\}$ . We wish to estimate the PSD function  $S(f)$  of  $\{X(n)\}$  from the data record. The plot below shows a basic estimate of  $S(f)$  referred to as the periodogram and denoted as  $I_N(f)$ . Also the true  $S(f)$  is depicted.



- Explain why  $I_N(f)$  reveals such high oscillations. Quantify this in terms of the bias and variance of  $I_N(f)$ . Is  $I_N(f)$  converging to  $S(f)$  as  $N \rightarrow \infty$  ?
- Introduce the concept of windowed periodogram estimates and explain why this class of estimates can converge to the true  $S(f)$ . In particular, elaborate on the bias-variance trade-off characterizing this class of estimates. Give an example, e.g., Bartlett window, of the proper window function yielding a consistent estimate of  $S(f)$ .
- The following plot shows various estimates  $\hat{S}_N(f)$  of  $S(f)$  based on  $N = 1024$  data points. It is assumed that the data were generated from the following sinusoidal signal model

$$X(n) = A_1 e^{j(2\pi f_1 n + \theta_1)} + A_2 e^{j(2\pi f_2 n + \theta_2)} + w(n),$$

where  $A_1, A_2, \theta_1, \theta_2$  are *iid* random variables uniformly distributed on  $[0, 2\pi]$  and  $w(n)$  is a zero-mean, white noise process. Based on this plot can you estimate the values of the unknown frequencies  $f_1$  and  $f_2$  ?

